

## Sheet Resistance

- Rewrite the resistance equation to separate  $(L/W)$ , the length-to-width ratio ... which is the number of “squares”  $N_{\square}$  from  $R_{\square}$ , the sheet resistance  $= (\sigma_n t)^{-1}$

$$R = \frac{L}{q\mu_n N_d W t} = \left( \frac{1}{q\mu_n N_d t} \right) \frac{L}{W} = R_{\square} (L/W) = R_{\square} N_{\square}$$

The sheet resistance is under the control of the *process designer*; the number of squares is determined by the layout and is specified by the *IC designer*.

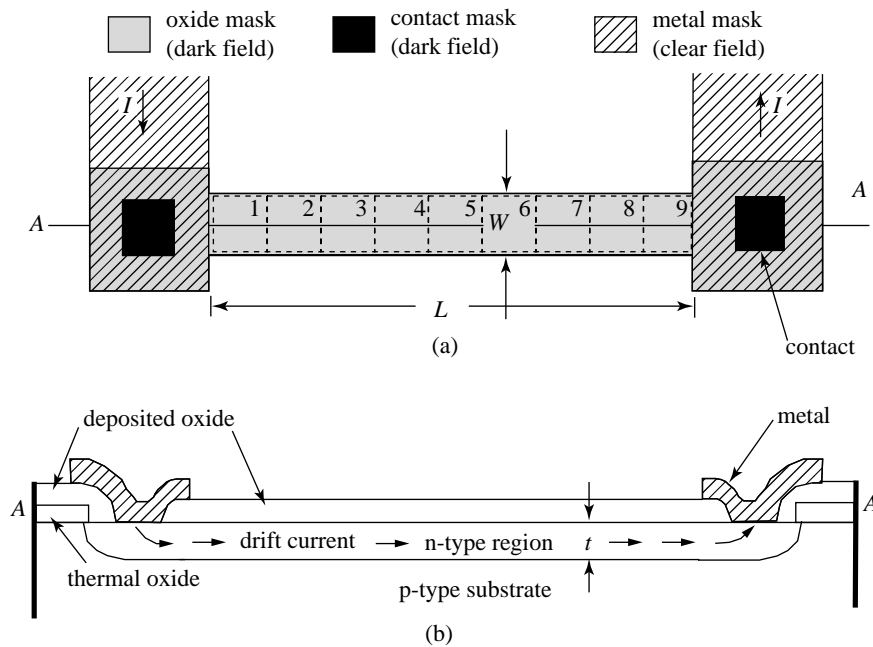
For average doping levels of  $10^{15} \text{ cm}^{-3}$  to  $10^{19} \text{ cm}^{-3}$  and a typical layer thickness of  $0.5 \text{ }\mu\text{m}$ , the sheet resistance ranges from  $100 \text{ k}\Omega/\square$  to  $10 \text{ }\Omega/\square$ .

Other conducting materials: (MOSIS  $1 \text{ }\mu\text{m}$  CMOS process)

	$\Omega / \square$
$n^+$ polysilicon ( $t = 500 \text{ nm}$ )	20
aluminum ( $t = 1 \text{ }\mu\text{m}$ )	0.07
silicided polysilicon	5
silicided source/drain diffusion	3

# Integrated Circuit Resistors

- Fabricate an n-type resistor in a p-type substrate using the process described in Chapter 2.



- Given the sheet resistance, we need to find the number of squares for this layout  
 $L / W = 9$  squares

# Laying Out a Resistor

- Rough approach:

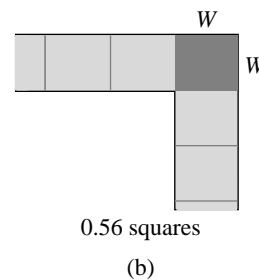
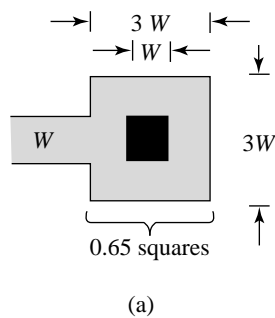
$$R \text{ known } \rightarrow N_{\square} = R / R_{\square}.$$

Select a width  $W$  (possibly the minimum to save area)  $\rightarrow$   
the length  $L = W N_{\square}$  and make a rectangle  $L \times W$  in area

Add contact regions at the ends ... ignore their contribution to  $R$

- More careful approach:

account for the contact regions and also, for corners



Measurement shows that the effective number of squares of the “dogbone” style contact region is 0.65 and for a  $90^{\circ}$  corner is 0.56.

For the resistor with  $L / W = 9$ , the contact regions add a significant amount to the total square count:

$$N_{\square} = 9 + 2 (0.65) = 10.3$$

In design, the contact regions and the corners should be accounted for to accurately determine the layout needed to yield the desired resistance.

## Uncertainties in IC Fabrication

The precision of transistors and passive components fabricated using IC technology is surprisingly, *poor*!

Sources of variations:

- ion impant dose varies from point to point over the wafer and from wafer to wafer
  
- thicknesses of layers after annealing vary due to temperature variations across the wafer
  
- widths of regions vary systematically due to imperfect wafer flatness (leading to focus problems) and randomly due to raggedness in the photoresist edges after development
  
- etc., etc.

## Quantifying Variations in Device Parameters

We will write an uncertain parameter (such as the acceptor conc.) as:

$$N_a = \overline{N}_a (1 \pm \varepsilon_{N_a})$$

where  $\overline{N}_a$  is the *average doping* and  $\varepsilon_{N_a}$  is the *normalized uncertainty*

(Note - we've swept a lot of probability and statistics under the rug here)

As an example, an average acceptor concentration  $N_a = 10^{16} \text{ cm}^{-3}$  and a normalized uncertainty  $\varepsilon_{N_a} = 0.04$  means that the acceptor concentration ranges from

$$0.96 \times 10^{16} \text{ cm}^{-3} \text{ to } 1.04 \times 10^{16} \text{ cm}^{-3}$$

How do variations combine to determine the variation in an IC resistor?

- assume that the variations are *independent*
- assume that the normalized uncertainty of a function of several variables is the square root of the sum of the squares of the individual uncertainties

$$\varepsilon_T = \sqrt{\varepsilon_a^2 + \varepsilon_b^2 + \varepsilon_c^2}$$

(Note - this assumes a “normal” distribution (bell curve))

## IC Resistor Uncertainty

$$R = \left( \frac{1}{qN_d\mu_n t} \right) \left( \frac{L}{W} \right)$$

Note that  $N_d$ ,  $\mu_n$ ,  $t$ ,  $L$ , and  $W$  are all subject to random variations

The average resistance is found by substituting the averages:

$$\bar{R} = \left( \frac{1}{q\bar{N}_d\bar{\mu}_n\bar{t}} \right) \left( \frac{\bar{L}}{\bar{W}} \right)$$

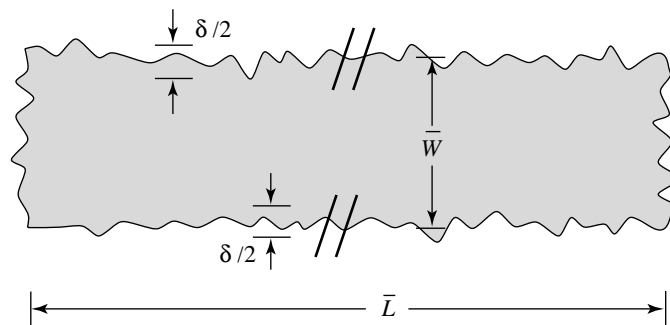
The normalized uncertainty in resistance is found from the “sum of squares” of the normalized uncertainties in  $N_d$ ,  $\mu_n$ ,  $t$ ,  $L$ , and  $W$

$$\varepsilon_R = \sqrt{\varepsilon_{N_d}^2 + \varepsilon_{\mu_n}^2 + \varepsilon_t^2 + \varepsilon_L^2 + \varepsilon_W^2}$$

This estimate is reasonable for relatively small uncertainties  $\varepsilon < 0.1$  and independent variables

## Linewidth Uncertainties

- Due to lithographic and etching variation, the edges of a rectangle are “ragged” -  
- greatly exaggerated in the figure



- The width is

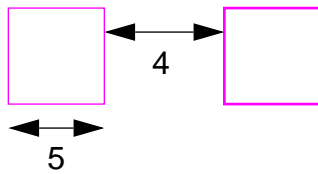
$$W = \bar{W} \pm \frac{\delta}{2} \pm \frac{\delta}{2} = \bar{W} \pm \delta \quad \rightarrow \quad W = \bar{W} \left( 1 \pm \frac{\delta}{\bar{W}} \right) = \bar{W} (1 \pm \epsilon_W)$$

- Conclusion 1: *wider* resistors have smaller normalized uncertainty (since  $\delta$  is independent of width)
- Conclusion 2: the length  $L \gg W$  and so its normalized uncertainty is negligible compared to that of  $W$

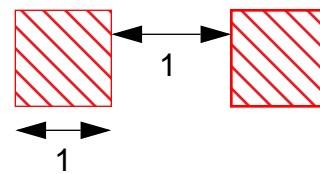
# Geometric Design Rules

Uncertainties in the linewidth and the overlay precision of successive masks determines (in part) the rules for laying out masks

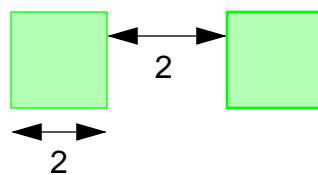
n-well



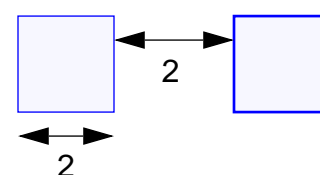
polysilicon



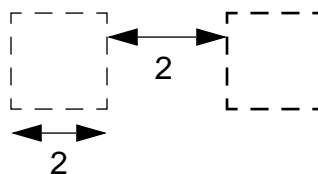
active



metal



select



contact

